# An 8.62 µW Processor for Autism Spectrum Disorder Classification using Shallow Neural Network

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Abstract— Autism Spectrum Disorder (ASD) is the prevalent child neurological and developmental disorder causing cognitive and behavioral impairments. The early diagnosis is an urgent need for treatment and rehabilitation of ASD patients. This work presents an electroencephalogram (EEG) based ASD classification processor that targets a patchform factor sensor that can be used for long time monitoring in a wearable environment. The selection of frontal and parietal lobe electrodes causes minimum uneasiness to the children. The proposed and implemented algorithm utilizes only four EEG electrodes. The processor is implemented and validated on Artix-7 FPGA which requires only 26K lookup tables and 15K flip flops. The hardware efficient implementation of the complex kurtosis value and Katz fractal dimension (KFD) features using kurtosis value indicator and KFD indicator with 54% and 38% efficient implementations, respectively, is provided. A hardware feasible shallow neural network architecture is used for the ASD classification. The system classifies the ASD with a high classification accuracy of 85.5% using the power and latency of 8.62µW and 2.25ms, respectively.

Keywords— Autism, neural network, neurological disorder, processor, wearable devices

## I. INTRODUCTION

Autism spectrum disorder (ASD) is a wide spectrum of neurological disorders including genetic and non-genetic factors. The term "spectrum" represents the widespread series of impairments associated with the disorder, causing the early diagnosis challenging. The recent estimates show a significant rise in the number of ASD patients across all ethnicities and socio-economic ranks [1]. The ASD diagnosis is standardly performed by Autism diagnostic observation schedule, 2nd Edition (ADOS-2) [1], requiring extensive and frequent behavioural observations leading to late diagnosis [1]. The ADOS-2 evaluations take ample time and may be avoided by many parents due to the feeling of disgrace and repeated visits to neurologists. The proposed solution would be able to diagnose a child as ASD or typical developing (TD) earlier by wearing a comfortable head-band with an integrated The processor would pre-process processor. Electroencephalogram (EEG), extract suitable features, and classify the child as ASD using suitable machine learning (ML) or deep neural network (DNN) classification. It would not only avoid the stigma associated with the prolonged diagnosis but also reduce the rehabilitation costs due to early intervention [2].

One of the biggest challenges in ASD detection at an early stage is the uncooperative behaviour of ASD children, therefore we are proposing to develop a miniaturized wearable

device to record and process EEG data. Only transmitting the EEG data wirelessly for remote processing will consume >15mW power, which is not suitable for children under the age of 4 years [3], [4]. Hence, a fully on-chip low-power system ( $\sim 0.5 mW$ ) will be necessitated to extract the features and classify the ASD from EEG data on the sensor to assist the neurologist in early detection.

EEG signals record the electrical activity inside the human brain using a certain number of electrodes. Despite the various challenges related to EEG signal acquisition, there is significant research to show the effectivity of scalp EEG for ASD [5]. There are some solutions to assist ASD children using their emotions [6], [7]. But no hardware-based ASD prediction processor is available. This paper presents the first low-power processor to classify ASD patients using a dataset recorded, trained, and tested on ADOS-2 confirmed ASD patients.

# II. METHODS AND TECHNIQUES

The ASD prediction using the EEG signal involves the acquisition of EEG data, EEG signal pre-processing for noise and artifacts removal, suitable feature extraction, and ML/deep learning classification. Fig. 1 shows the top-level diagram of our ASD classification processor. The upper portion shows the offline analysis carried out in different python packages to identify the suitable features and channels required for ASD classification [8]. The selected feature-set and channel list were further optimized for 1) the hardware realization and 2) a shallow neural network (SNN) classifier. Neural networks are capable of learning small datasets quickly [9]. The SNN architecture presented in this paper provided us better classification results with minimum hardware resources against other ML models. The SNN parameters (weights and biases) were then uploaded to our ASD classification processor for online testing and verification. The bottom part of the figure shows our hardware-based ASD classification processor including the EEG pre-processing unit, feature extraction unit, and SNN classifier. The processor classifies a subject as ASD or TD using the EEG signals of the selected four channels.

## A. Dataset

To develop an efficient algorithm for the ASD classification, we have utilized the data recorded by ref [5]. The data provides the EEG dataset of 17 participants including 8 ASD patients and 9 TD subjects using 32 electrodes. The EEG data were sampled at 250 Hz for 9 minutes duration.

### B. Channels Selection

The selection of a limited number of channels and their scalp locations are quite important for the continuous monitoring of EEG data of ASD patients [6].

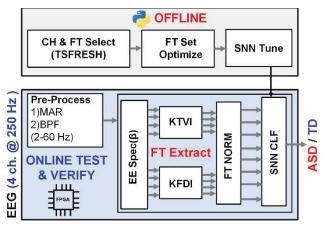


Fig. 1. Top-level block diagram of the ASD classification processor

Due to the discomfort involved and the hardware infeasibility, a large number of EEG channels and bulky EEG headsets are not suitable for a wearable device [10][11]. The initial analysis of the EEG signals for ASD and TD subjects identified different channels and features to be quite important for ASD classification. A four-channel set (F7, F8, CP5, CP2) was chosen for our ASD classification processor. These channels differentiate ASD and TD children using frontal and central parietal connectivity differences. The selected channels classified a subject as an ASD or TD with 85.5% classification accuracy with selected features and classification algorithm.

# C. Feature Extraction Engine

The feature extraction requires the identified features of Kurtosis Value (KTV) and Katz Fractal Dimension (KFD) in the beta (12-30 Hz) frequency band. KTV provides information about the degree of concentration of the signal around the mean [12]. KFD provides information about the energy decay of a signal [13]. Eq. (1)-(2) define KTV and KFD. The 10-bits digitized and pre-processed (P.Process) EEG data sampled at 250 Hz is forwarded to the feature extraction engine (FEE). The FEE passes the EEG signal from a bandpass filter and then calculates the KTV or KFD feature using the required hardware components. F<sub>x</sub> represents the or KFD feature for a single EEG channel.

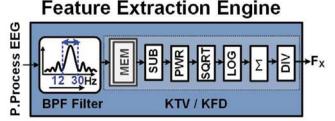


Fig. 2. Feature Extraction Engine highlighting a single channel

$$KTV = \sum_{i=0}^{n} \frac{(x_{i} - \bar{X})^{4}}{(N-1)*S^{4}}$$
 (1)

$$KFD = \frac{\log(\sum ED(X_i, X_{i-1}))}{\log(\max(\operatorname{ecd}(X_i, X_{i-1})))}$$
 (2)

 $X_i$ ,  $\bar{X}$ , N, and S represent the time-series EEG sample, mean EEG value, total number of EEG samples, and standard deviation of the EEG data, respectively. KTV (1) calculates the ratio of the fourth power summation of differences of X<sub>i</sub> and  $\bar{X}$  with a product of one less than the total number of samples (N-1) and the fourth power of S. KFD (2) calculates the ratio between logarithms of summation of Euclidean differences (ED) between consecutive EEG samples (Xi and X<sub>i-1</sub>) and the maximum ED. The calculation of these features requires huge memory requirements (> 15 MB) along with complex floating-point (FP) logarithm, power, and square root calculations (1)-(2). These calculations would make the ASD classification processor's hardware implementation unrealizable and impractical due to high power consumption (> 500 mW) and huge silicon area requirements or FPGA resource constraints. Therefore, it is quite important to optimize these features to a hardware realizable approximation.

$$KTVI = K * SDI^4$$
 (3)

$$SDI = \frac{\max(X) - \min(X)}{4} \tag{4}$$

$$SDI = \frac{\max(X) - \min(X)}{4}$$

$$KFDI = \max(X_i - X_{i-1})^2 - \sum_{i=0}^{n} (X_i - X_{i-1})^2$$
 (5)

KTV (1) and KFD (2) were approximated to KTV indication (KTVI) and KFD indication (KFDI), respectively. KTVI (3) was calculated using the product of the fourth power of standard deviation indicator (SDI) and a constant parameter K. SDI is the approximated standard deviation using range rule [14]. The SDI simply requires the difference between the maximum and minimum samples in the EEG time series represented by max(X) and min(X), respectively. Eq (4) shows the calculation of SDI where X represents the EEG data. The KFDI (5) calculates the difference between squares of the maximum difference and the total difference between  $X_i$  and  $X_{i-1}$ .  $X_i$  and  $X_{i-1}$  represent the current and previous EEG

Fig. 3 shows the FEE to calculate the KTVI. The preprocessed EEG data in the beta frequency band (EEG β band) and the electrode/ channel number (ELT) were forwarded as input to the KTVI calculation unit. The EEG  $\beta$  band was calculated using a quantized FIR filter of 30th order as a halfprecision (16'b) FP value.

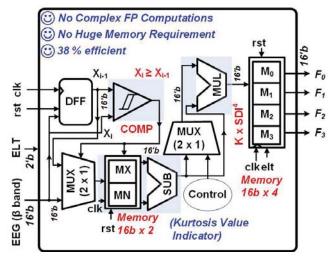


Fig. 3. FEE highlighting KTVI

The ELT represents the current electrode from the subset of four electrodes used for the classification. A comparator unit (COMP) compares the consecutive EEG samples and raises the output flag if the current EEG sample (Xi) is higher than the previous sample  $(X_{i-1})$ . The output flag of the comparator is used as a section input for a two-to-one multiplexer to update the contents of minimum (MN) and maximum (MX) values. Xi and Xi-1 were sampled by a flip-flop (DFF) and forwarded as inputs to the multiplexer. A 32-bits memory unit block (16'b x 2) was used to store the MN and MX values. An FP subtractor (SUB) calculates the difference between MX and MN values. The fourth power SDI and KTVI were calculated using a single FP multiplication unit (MUL) controlled by the control unit. The KTVI of the selected channel was stored in a memory block (16'b x 4) using ELT.  $F_{0-3}$  represents the KTVI of the selected four channels. The proposed KTVI implementation does not require any complex FP calculations and huge memory requirements and was 38% efficient than conventional KTV implementation using (1).

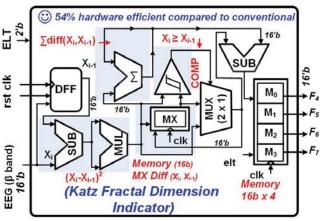


Fig. 4. Feature Extraction Engine highlighting KFDI

Fig. 4 shows the FEE to calculate the KFDI. The KFDI similarly requires the EEG β band and ELT. Xi and Xi-1 were sampled using a DFF and the difference between Xi and Xi-1 was calculated similarly to KTVI. A 16-bits FP summation unit  $(\Sigma)$  was used to calculate the summation of differences (5) between Xi and Xi-1. The maximum difference (MX Diff (Xi, Xi-1)) was calculated using a FP comparator (COMP) controlled by a 2-1 multiplexer (MUX). MX was stored in a 16-bits memory block and updated using a MUX controlled by the COMP. A FP subtraction unit (SUB) and multiplication unit (MUL) were used to calculate the squared difference (5) between Xi and Xi-1. The KFDI of the selected channel was similarly stored in a memory block (16'b x 4) using ELT. F4, F5, F6, and F8 represent the KFDI of the selected four channels. The proposed KFDI implementation does not require any complex FP calculations and huge memory requirements and was 54% efficient than conventional (2) KFD implementation. The calculated features F0-7 were forwarded to the SNN classification unit after normalization as a feature vector.

### D. Shallow Neural Network Classification Unit

An SNN is a fully connected neural network without multiple hidden layers. The SNN classifies the output as ASD and TD by adjusting or optimizing the weights and biases during the learning process from the difference between the desired and the actual output through backpropagation. The input layer (6) calculated the hidden layer values  $(N_{0-49})$  using multiplications and additions with the parameters  $(P_{0-449})$  and a sigmoid function. The output layer (7) values  $(O_{0-1})$  are calculated using  $N_{0-449}$  and output layer parameters  $(P_{450-551})$ . The higher value of  $O_0$  or  $O_1$  classifies the patient as ASD or TD, respectively.

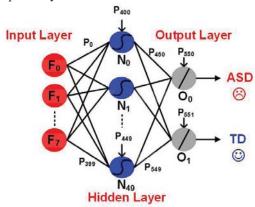


Fig. 5. Shallow Neural Network architecture

The higher value of  $O_0$  or  $O_1$  classifies the patient as ASD or TD, respectively. Fig. 5 shows the architecture of the SNN used for ASD classification. The SNN contains eight, fifty, and two nodes in the input, hidden, and output layers, respectively. The eight normalized features ( $F_0$ - $F_7$ ) are forwarded to the input layer.

$$N_{0-49} = Sigmoid(\sum_{a,b=0,c=400}^{a=399,b=7,c=449} P_a.F_b + P_c)$$
 (6)

$$O_{0-1} = \sum_{a=450, b=0, c=550}^{a=459, b=49, c=551} (P_a. N_b + P_c)$$
 (7)

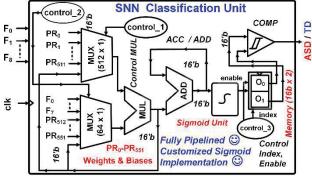


Fig. 6. Proposed SNN classification unit architecture

Eq. (6)-(7) represents the mathematical operations required for SNN implementation. P<sub>0-399</sub> and P<sub>400-449</sub> are the weights and biases for the input layer, respectively. P<sub>450-459</sub> and P<sub>550-551</sub> are the weights and biases for the output layer, respectively. Fig. 6 shows the hardware implementation of the SNN classification unit. The normalized features F<sub>0-7</sub> are inputted to the classification unit, which uses a FP multiplier and adder to perform the addition or accumulation functions (6)-(7). Two multiplexers (512 x 1 and 64 x 1) are used to select multiplier inputs to perform the multiplication, accumulation, or addition functions. Two finite state machine control units (control\_1 and control\_2) are used to provide the selection inputs of the multiplexers. A sigmoid unit is used to apply the sigmoid

activation function [15]. A 32-bits memory block is used to store  $O_0$  and  $O_1$ . A finite state machine control unit (control\_3) is used to control the memory block using index and enable. The classification output (ASD/TD) is calculated after comparing  $O_0$  and  $O_1$  using a FP comparator.

#### III. RESULTS AND DISCUSSION

The proposed ASD classification processor is implemented on Xilinx Atrix-7 FPGA. This work is the first hardware-based implementation of an ASD classification processor to the best of our knowledge verified on a dataset of ASD patients. The EEG dataset for ASD classification by [5] was used for this work. The overall power of 8.62  $\mu W$  is consumed while operating at 100 MHz clock. The processor classifies a patient as ASD or TD with 85.5% classification accuracy. The classification results were evaluated using a 5-fold cross-validation scheme.

Table 1. Comparison with the state-of-the-art

	IEEE ISCAS'19	IEEE TCASII'15	IEEE Access' 20	Nature SR' 18	IEEE IRI' 19	This Work
H/W	Yes (FPGA)	Yes (SoC)	Yes (FPGA)	No	No	Yes (FPGA)
Power	12.3 uW	13.6 uW	150 mW	NA	NA	8.62 uW
Electrodes Count	8 *	16 *	14 *	19	32	4
Accuracy	63 % *	100 % *	83.1 % *	95 %	95.5 %	85.5 %
LUTs		L	26229	NA	NA	18361
FFs	_		15180	NA	NA	10627
Application	Emotion	Epilespy	Emotion	ASD	ASD	ASD

The comparison of the work with previous ASD classification processors [5], [16] is shown in Table 1. Since no other hardware-based ASD classification processor exists, the results are also compared with similar systems for other biomedical applications [7], [17] - [19]. The classification performance of the work is quite good (85.5%) being the 1st hardware implementation and using the lowest number (4) of electrodes. Since the other hardware implementations target different biomedical applications, the classification accuracy does not represent a lateral comparison alone. The overall classification power, lookup tables, and flip-flops count are significantly lesser than epilepsy or emotion classification.

## IV. CONCLUSION

Wearable ASD classification processors can be a breakthrough in biomedical healthcare. They would assist ASD children and their caregivers in ASD diagnosis without any feeling of stigma. The implemented SNN classification processor utilizes the approximated and optimized implementations for hardware costly KTV and KFD features with 38% and 54% lesser hardware resources compared to conventional implementation. The high classification results and lower hardware resources are quite encouraging to develop a fully integrated system for ASD classification after validation of the system after incorporating more ASD datasets.

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#### REFERENCES

- [1] "Autism Speaks" [Online]. Available: https://www.autismspeaks.org/autism-statistics
- [2] E. Fuller and A. Kaiser, "The Efects of Early Intervention on Social Communication Outcomes for Children with Autism Spectrum Disorder: A Meta-analysis," *Journal of Autism and Developmental Disorders*, vol. 50, no. 1, pp. 1683-1700, May. 2020.
- [3] W. Saadeh, F. H. Khan and M. Altaf, "Design and Implementation of a Machine Learning Based EEG Processor for Accurate Estimation of Depth of Anesthesia," *IEEE Transactions on Biomedical Circuits and Systems*, vol. 13, no. 4, pp. 658-669, Aug. 2019.
- [4] F. H. Khan, U. Ashraf, M. Altaf and W. Saadeh, "A Patient-Specific Machine Learning based EEG Processor for Accurate Estimation of Depth of Anesthesia," *IEEE Biomedical Circuits and Systems Conference (BioCAS)*, Oct. 2018, pp. 1-4.
- [5] Y. Jayawardana, M. Jaime and S. Jayarathna, "Analysis of Temporal Relationships between ASD and Brain Activity through EEG and Machine Learning," in *IEEE 20th International Conference on Information Reuse and Integration for Data Science (IRI)*, Aug. 2019, p. 151-158.
- [6] A. Aslam and M. Altaf, "An On-Chip Processor for Chronic Neurological Disorders Assistance Using Negative Affectivity Classification," *IEEE Transactions on Biomedical Circuits and Systems*, vol. 14, no. 4, pp. 838-851, Aug. 2020.
- [7] A. Aslam, and M. Altaf, "An 8 Channel Patient-Specific Neuromorphic Processor for the Early Screening of Autistic Children through Emotion Detection," *IEEE International Symposium on Circuits and Systems (ISCAS)*, May. 2019, pp. 1-4.
- [8] M. Christ, N. Braun, J. Neuffer and A. Kempa-Liehr, "Time Series FeatuRe Extraction on basis of Scalable Hypothesis tests," *Neurocomputing*, vol. 307, pp. 72-77, Sept. 2018.
- [9] M. Olson, A. Wyner, and R. Berk, "Modern neural networks generalize on small data sets" in *Advances in Neural Information Processing Systems*, Red Hook, NY, USA:Curran Associates, pp. 3619-3628, Dec. 2018
- [10] S. A. Zamin, M. A. B. Altaf and W. Saadeh, "A Single Channel EEG-based All AASM Sleep Stages Classifier for Neurodegenerative Disorder," *IEEE Biomedical Circuits and Systems Conference (BioCAS)*, Oct. 2019, pp. 1-4.
- [11] M. A. B. Altaf and W. Saadeh, "A 0.21 µJ patient-specific REM/Non-REM sleep classifier for Alzheimer patients," *IEEE Biomedical Circuits and Systems Conference (BioCAS)*, Oct. 2017, pp. 1-4.
- [12] F. Al-Athari, "Confidence Interval for Locations of Non-kurtosis and Large Kurtosis Leptokurtic Symmetric Distributions", *Journal of Applied Sciences*, vol. 11, no. 3, pp. 528-534, 2011.
- [13] D. R. Jevtić and M. P. Paskaš, "Application of Katz algorithm for fractal dimension in analysis of room impulse response," in 19thTelecommunications Forum (TELFOR) Proceedings of Papers, November. 2011, pp. 1063-1066...
- [14] X. Wan, W. Wang, J. Liu and T. Tong, "Estimating the sample mean and standard deviation from the sample size, median, range and/or interquartile range," *BMC Medical Research Methodology*, vol. 14, no. 1, Dec. 2014.
- [15] A. Aslam, T. Iqbal, M. Aftab, W. Saadeh and M. Altaf, "A10.13uJ/classification 2-channel Deep Neural Network-based SoC for Emotion Detection of Autistic Children," in *IEEE Custom Integrated Circuits Conference (CICC)*, Mar. 2020, pp. 1-4.
- [16] W. Bosl, A. Tierney, H. Tager-Flusberg and C. Nelson, "EEG Analytics for Early Detection of Autism Spectrum Disorder: A datadriven approach," Scientific Reports, vol. 9, no. 6828, May. 2018.
- [17] M. Shoaran, C. Pollo, K. Schindler and A. Schmid, "A Fully Integrated IC With 0.85-μW/Channel Consumption for Epileptic iEEG Detection," *IEEE Transactions on Circuits and Systems II: Express Briefs*, vol. 62, no. 2, pp. 114-118, Feb. 2015.
- [18] H. A. Gonzalez, S. Muzaffar, J. Yoo and I. M. Elfadel, "BioCNN: A Hardware Inference Engine for EEG-Based Emotion Detection," *IEEE Access*, vol. 8, pp. 140896-140914, Jul. 2020.
- [19] M. R. Khan, W. Saadeh and M. A. B. Altaf, "A low complexity patient-specific threshold based accelerator for the Grand-mal seizure disorder," *IEEE Biomedical Circuits and Systems Conference (BioCAS)*, Oct. 2017, pp. 1-4.